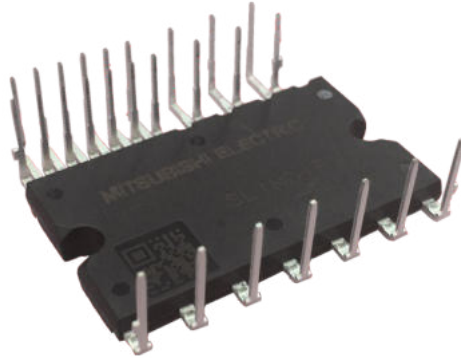


< DIIPM >

# SLIMDIP-M

TRANSFER MOLDING TYPE  
INSULATED TYPE

## OUTLINE



Normal terminal type

## MAIN FUNCTION AND RATINGS

- RC-IGBT inverter bridge for three phase DC-to-AC power conversion
- Built-in bootstrap diodes with current limiting resistor
- Open emitter type

## APPLICATION

- AC 100~240V (DC voltage:400V or below) three phase low power motor inverter drive

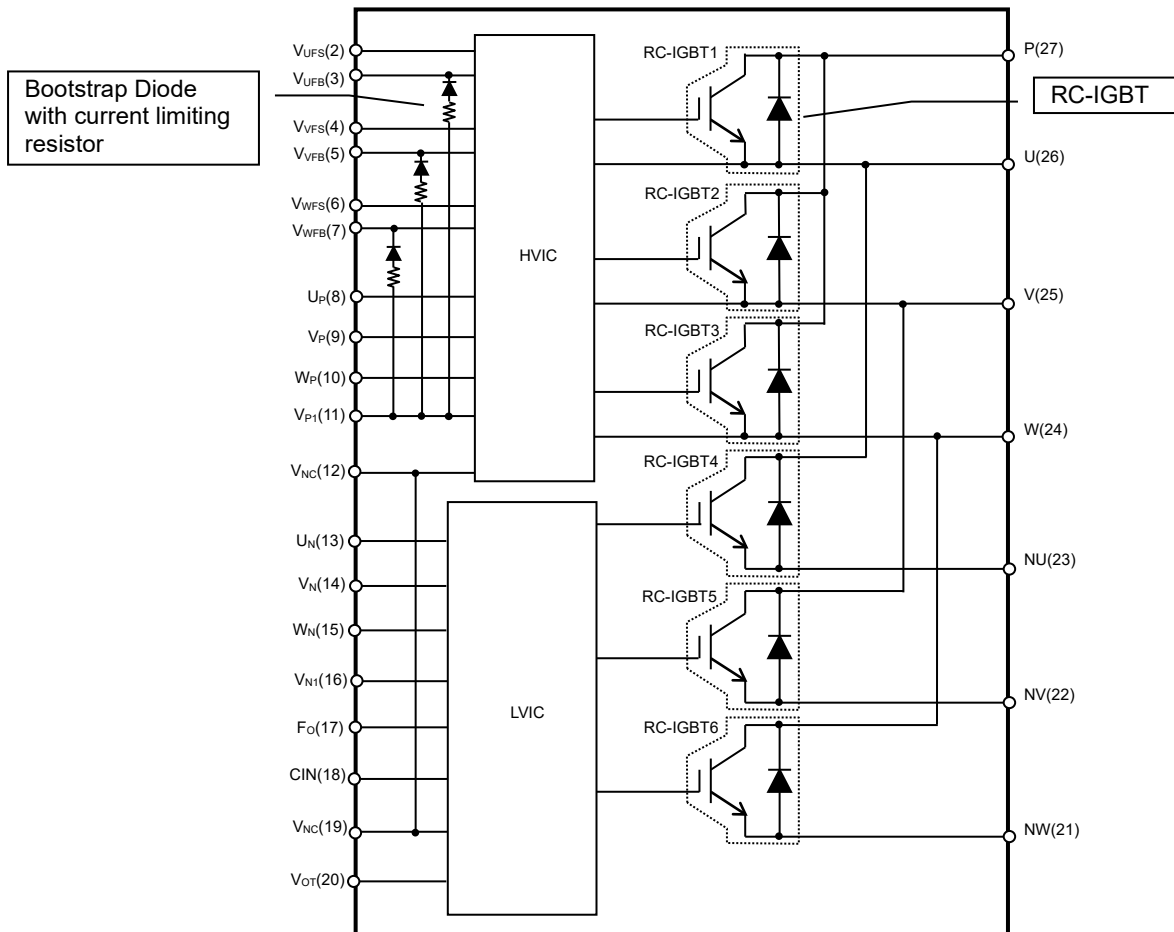
## TERMINAL LINE UP

| Terminal        | Part number | Suffix |
|-----------------|-------------|--------|
| Normal terminal | SLIMDIP-M   | 500    |
| Short terminal  | SLIMDIP-M   | 505    |

## INTEGRATED DRIVE, PROTECTION AND SYSTEM CONTROL FUNCTIONS

- For P-side : Drive circuit, High voltage high-speed level shifting, Control supply under-voltage protection (UV)
- For N-side : Drive circuit, Control supply under-voltage protection (UV), Short circuit protection (SC), Over temperature protection (OT)
- Fault signaling : Corresponding to SC fault (N-side IGBT), UV fault (N-side supply) and OT fault
- Temperature monitoring : Outputting LVIC temperature by analog signal
- Input interface : Schmitt-triggered 3V, 5V input compatible, high active logic.
- UL Recognized : UL1557 File E323585

## INTERNAL CIRCUIT



**SLIMDIP-M****TRANSFER MOLDING TYPE  
INSULATED TYPE****MAXIMUM RATINGS** ( $T_j = 25^\circ\text{C}$ , unless otherwise noted)**INVERTER PART**

| Symbol                 | Parameter                          | Condition                                                              | Ratings  | Unit             |
|------------------------|------------------------------------|------------------------------------------------------------------------|----------|------------------|
| $V_{CC}$               | Supply voltage                     | Applied between P-NU, NV, NW                                           | 450      | V                |
| $V_{CC(\text{surge})}$ | Supply voltage (surge)             | Applied between P-NU, NV, NW                                           | 500      | V                |
| $V_{CES}$              | Collector-emitter voltage          |                                                                        | 600      | V                |
| $\pm I_C$              | Each IGBT collector current        | $T_C = 25^\circ\text{C}$ (Note 1)                                      | 10       | A                |
| $\pm I_{CP}$           | Each IGBT collector current (peak) | $T_C = 25^\circ\text{C}$ , less than 1ms<br>$V_D, V_{DB} = 15\text{V}$ | 20       | A                |
| $T_j$                  | Junction temperature               | (Note 2)                                                               | -30~+150 | $^\circ\text{C}$ |

Note1: Pulse width and period are limited due to junction temperature.

Note2: The maximum junction temperature rating of built-in chips is  $150^\circ\text{C}$  ( $@T_C \leq 115^\circ\text{C}$ ). However, to ensure safe operation of DIIPM, the average junction temperature should be limited to  $T_{j(\text{Ave})} \leq 125^\circ\text{C}$  ( $@T_C \leq 115^\circ\text{C}$ ).

**CONTROL (PROTECTION) PART**

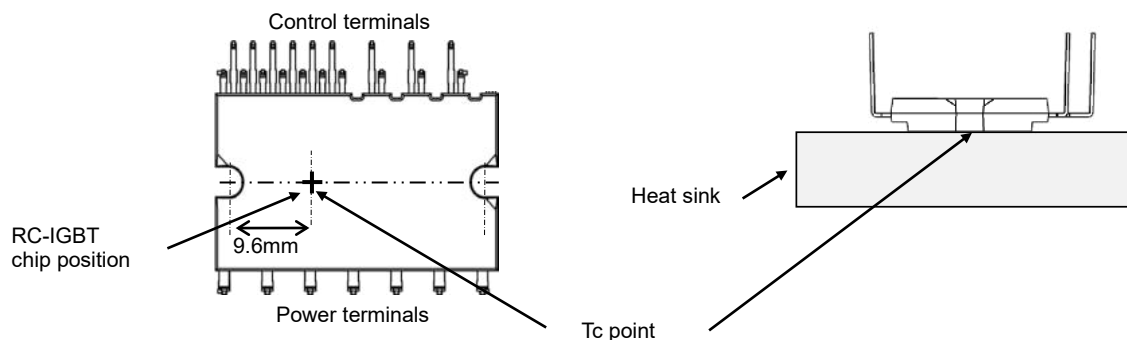
| Symbol   | Parameter                     | Condition                                                                 | Ratings         | Unit |
|----------|-------------------------------|---------------------------------------------------------------------------|-----------------|------|
| $V_D$    | Control supply voltage        | Applied between $V_{P1}-V_{NC}$ , $V_{N1}-V_{NC}$                         | 20              | V    |
| $V_{DB}$ | Control supply voltage        | Applied between $V_{UFB}-V_{UFS}$ , $V_{VFB}-V_{VFS}$ , $V_{WFB}-V_{WFS}$ | 20              | V    |
| $V_{IN}$ | Input voltage                 | Applied between $U_P$ , $V_P$ , $W_P$ , $U_N$ , $V_N$ , $W_N-V_{NC}$      | -0.5~ $V_D+0.5$ | V    |
| $V_{FO}$ | Fault output supply voltage   | Applied between $F_O-V_{NC}$                                              | -0.5~ $V_D+0.5$ | V    |
| $I_{FO}$ | Fault output current          | Sink current at $F_O$ terminal                                            | 1               | mA   |
| $V_{SC}$ | Current sensing input voltage | Applied between $CIN-V_{NC}$                                              | -0.5~ $V_D+0.5$ | V    |

**TOTAL SYSTEM**

| Symbol                | Parameter                                                                     | Condition                                                                                                              | Ratings  | Unit             |
|-----------------------|-------------------------------------------------------------------------------|------------------------------------------------------------------------------------------------------------------------|----------|------------------|
| $V_{CC(\text{PROT})}$ | Self protection supply voltage limit<br>(Short circuit protection capability) | $V_D = 13.5\sim 16.5\text{V}$ , Inverter Part<br>$T_j = 150^\circ\text{C}$ , non-repetitive, less than 2 $\mu\text{s}$ | 400      | V                |
| $T_C$                 | Module case operation temperature                                             | Measurement point of $T_C$ is described in Fig.1 (Note3)                                                               | -30~+115 | $^\circ\text{C}$ |
| $T_{stg}$             | Storage temperature                                                           |                                                                                                                        | -40~+125 | $^\circ\text{C}$ |
| $V_{iso}$             | Isolation voltage                                                             | 60Hz, Sinusoidal, AC 1min, between connected all pins<br>and heat sink plate                                           | 2000     | $V_{rms}$        |

Note3  $T_C$  MEASUREMENT POINT

Fig. 1

**THERMAL RESISTANCE**

| Symbol          | Parameter                                       | Condition                              | Limits |      |      | Unit |
|-----------------|-------------------------------------------------|----------------------------------------|--------|------|------|------|
|                 |                                                 |                                        | Min.   | Typ. | Max. |      |
| $R_{th(j-c)/Q}$ | Junction to case thermal resistance<br>(Note 4) | Inverter RC-IGBT part (per 1/6 module) | -      | -    | 5.2  | K/W  |

Note 4: Grease with good thermal conductivity and long-term endurance should be applied evenly with about  $+100\mu\text{m} \sim +200\mu\text{m}$  on the contacting surface of DIIPM and heat sink. The contacting thermal resistance between DIIPM case and heat sink  $R_{th(c-f)}$  is determined by the thickness and the thermal conductivity of the applied grease. For reference,  $R_{th(c-f)}$  is about 0.4K/W (per 1/6 module, grease thickness: 20 $\mu\text{m}$ , thermal conductivity: 1.0W/m $\cdot$ K).

**ELECTRICAL CHARACTERISTICS** ( $T_J = 25^\circ\text{C}$ , unless otherwise noted)**INVERTER PART**

| Symbol        | Parameter                            | Condition                                                                                                                         |                              | Limits |      |      | Unit    |
|---------------|--------------------------------------|-----------------------------------------------------------------------------------------------------------------------------------|------------------------------|--------|------|------|---------|
|               |                                      |                                                                                                                                   |                              | Min.   | Typ. | Max. |         |
| $V_{CE(sat)}$ | Collector-emitter saturation voltage | $V_D=V_{DB} = 15V, V_{IN}= 5V$                                                                                                    | $I_C= 10A, T_J= 25^{\circ}C$ | -      | 2.25 | 2.85 | V       |
| $V_{EC}$      | FWDi forward voltage                 | $V_{IN}= 0V, -I_C= 10A$                                                                                                           |                              | -      | 1.90 | 2.40 | V       |
| $t_{on}$      | Switching times                      | $V_{CC}= 300V, V_D= V_{DB}= 15V$<br>$I_C= 10A, T_J= 25^{\circ}C, V_{IN}= 0\leftrightarrow 5V$<br>Inductive Load (upper-lower arm) |                              | 0.50   | 0.85 | 1.20 | $\mu s$ |
| $t_{C(on)}$   |                                      |                                                                                                                                   |                              | -      | 0.30 | 0.50 | $\mu s$ |
| $t_{off}$     |                                      |                                                                                                                                   |                              | -      | 0.85 | 1.20 | $\mu s$ |
| $t_{C(off)}$  |                                      |                                                                                                                                   |                              | -      | 0.10 | 0.30 | $\mu s$ |
| $t_{rr}$      |                                      |                                                                                                                                   |                              | -      | 0.10 | -    | $\mu s$ |
| $I_{CES}$     | Collector-emitter cut-off current    | $V_{CE}=V_{CES}$                                                                                                                  | $T_J= 25^{\circ}C$           | -      | -    | 1    | mA      |

**CONTROL (PROTECTION) PART**

| Symbol        | Parameter                                          | Condition                                                                             |                                 | Limits                   |       |       | Unit        |
|---------------|----------------------------------------------------|---------------------------------------------------------------------------------------|---------------------------------|--------------------------|-------|-------|-------------|
|               |                                                    |                                                                                       |                                 | Min.                     | Typ.  | Max.  |             |
| $I_D$         | Circuit current                                    | Total of $V_{P1}$ - $V_{NC}$ , $V_{N1}$ - $V_{NC}$                                    | $V_D=15V$ , $V_{IN}=0V$         | -                        | -     | 3.10  | mA          |
|               |                                                    |                                                                                       | $V_D=15V$ , $V_{IN}=5V$         | -                        | -     | 4.20  |             |
| $I_{DB}$      |                                                    | Each part of $V_{UFB}$ - $V_{UFS}$ ,<br>$V_{VFB}$ - $V_{VFS}$ , $V_{WFB}$ - $V_{WFS}$ | $V_D=V_{DB}=15V$ , $V_{IN}=0V$  | -                        | -     | 0.10  |             |
|               |                                                    |                                                                                       | $V_D=V_{DB}=15V$ , $V_{IN}=5V$  | -                        | -     | 0.10  |             |
| $V_{SC(ref)}$ | Short circuit trip level                           | $V_D = 15V$ (Note 5)                                                                  |                                 | 0.455                    | 0.480 | 0.505 | V           |
| $UV_{DBt}$    | P-side Control supply under-voltage protection(UV) | Trip level                                                                            |                                 | 9.0                      | 10.0  | 12.0  | V           |
| $UV_{DBr}$    |                                                    | Reset level                                                                           |                                 | 9.0                      | 10.0  | 12.0  | V           |
| $UV_{Dt}$     | N-side Control supply under-voltage protection(UV) | Trip level                                                                            |                                 | 10.3                     | -     | 12.5  | V           |
| $UV_{Dr}$     |                                                    | Reset level                                                                           |                                 | 10.8                     | -     | 13.0  | V           |
| $V_{OT}$      | Temperature Output                                 | Pull down R=5.1k $\Omega$ (Note 6)                                                    | LVIC Temperature=95 $^{\circ}C$ | 2.76                     | 2.89  | 3.03  | V           |
| $OT_t$        | Over temperature protection (Note 7)               | $V_D = 15V$ ,<br>Detect LVIC temperature                                              | Trip level                      | 115                      | 130   | 145   | $^{\circ}C$ |
| $OT_{rh}$     |                                                    |                                                                                       |                                 | Hysteresis of trip-reset | -     | 10    | -           |
| $V_{FOH}$     | Fault output voltage                               | $V_{SC} = 0V$ , $F_O$ terminal pulled up to 5V by 10k $\Omega$                        |                                 | 4.9                      | -     | -     | V           |
| $V_{FOL}$     |                                                    | $V_{SC} = 1V$ , $I_{FO} = 1mA$                                                        |                                 | -                        | -     | 0.95  | V           |
| $t_{FO}$      | Fault output pulse width                           | (Note 8)                                                                              |                                 | 20                       | -     | -     | $\mu s$     |
| $I_{IN}$      | Input current                                      | $V_{IN} = 5V$                                                                         |                                 | 0.70                     | 1.00  | 1.50  | mA          |
| $V_{th(on)}$  | ON threshold voltage                               | Applied between $U_P$ , $V_P$ , $W_P$ , $U_N$ , $V_N$ , $W_N$ - $V_{NC}$              |                                 | -                        | 1.70  | 2.35  | V           |
| $V_{th(off)}$ | OFF threshold voltage                              |                                                                                       |                                 | 0.70                     | 1.30  | -     |             |
| $V_{th(hys)}$ | ON/OFF threshold hysteresis voltage                |                                                                                       |                                 | 0.25                     | 0.40  | -     |             |
| $V_F$         | Bootstrap Di forward voltage                       | $I_F$ =10mA including voltage drop by limiting resistor (Note 9)                      |                                 | 1.1                      | 1.7   | 2.3   | V           |
| R             | Built-in limiting resistance                       | Included in bootstrap Di                                                              |                                 | 80                       | 100   | 120   | $\Omega$    |

Note 5 : SC protection works only for N-side IGBT. Please select the external shunt resistance such that the SC trip-level is less than 1.5 times of the current rating in case of  $V_D < 15\text{V}$  and less than 1.7 times of the current rating in case of  $V_D \geq 15\text{V}$ .

6 : Temperature of LVIC vs.  $V_{OT}$  output characteristics is described in Fig.3. These minimum and maximum curves are based on theoretical designed value excluding LVIC temperature=95 $^\circ\text{C}$  limits.

7 : When the LVIC temperature exceeds OT trip temperature level( $OT_t$ ), OT protection works and  $F_O$  outputs. In that case if the heatsink dropped off or fixed loosely, don't reuse that DIIPM. (There is a possibility that junction temperature of power chips exceeded maximum  $T_J(150^\circ\text{C})$ .)

8 : Fault signal  $F_O$  outputs when SC, UV or OT protection works.  $F_O$  pulse width is different for each protection modes. At SC failure,  $F_O$  pulse width is a fixed width (=minimum 20 $\mu\text{s}$ ), but at UV or OT failure,  $F_O$  outputs continuously until recovering from UV or OT state. (But minimum  $F_O$  pulse width is 20 $\mu\text{s}$ .)

9 : The characteristics of bootstrap Di is described in Fig.2.

Fig. 2 Characteristics of Bootstrap Di  $V_F$ - $I_F$  curve (@ $T_a=25^\circ\text{C}$ ) Including Voltage Drop by Limiting Resistor (Right chart is enlarged chart.)

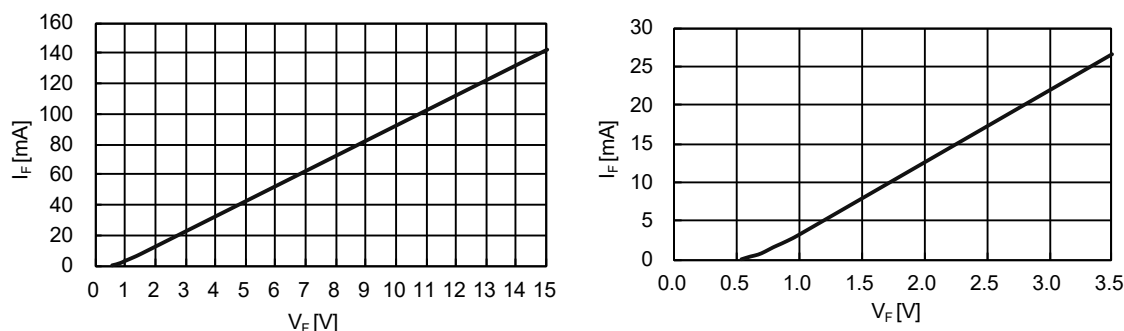
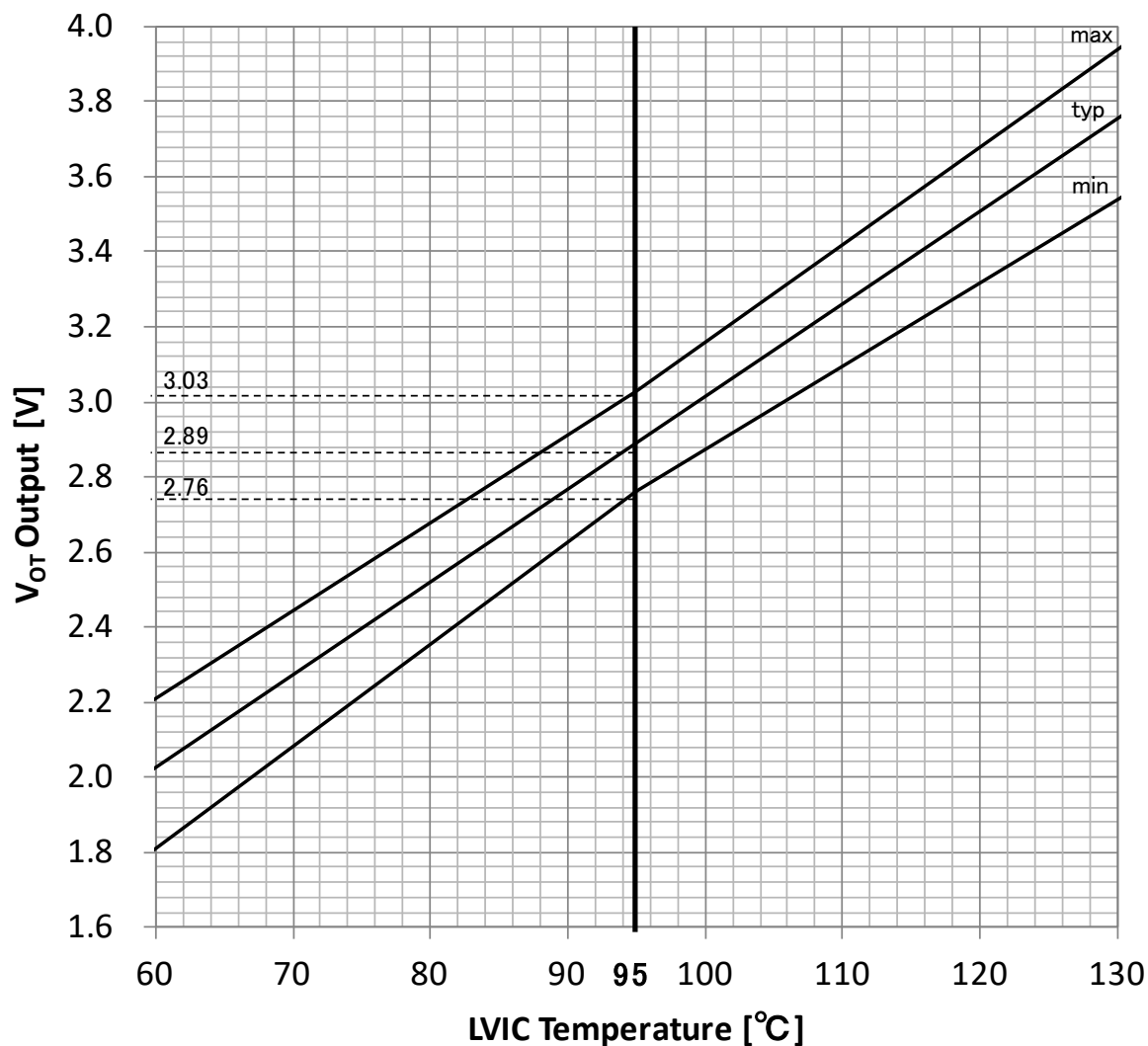
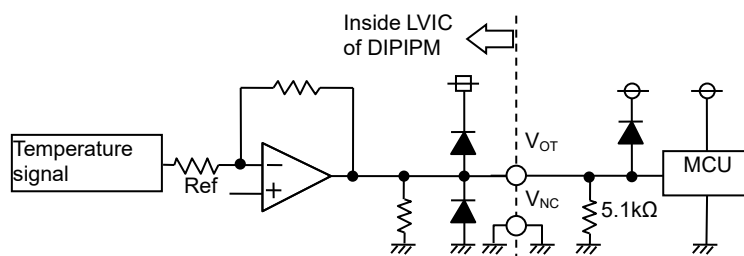


Fig. 3 Temperature of LVIC vs.  $V_{OT}$  Output CharacteristicsFig. 4 Pattern Wiring Around the Analog Voltage Output Circuit [ $V_{OT}$  terminal]

- (1)  $V_{OT}$  outputs the analog signal that is amplified signal of temperature detecting element on LVIC by inverting amplifier.
- (2) It is recommended to insert 5kΩ (5.1kΩ is recommended) pull down resistor for getting linear output characteristics at low temperature below room temperature. When the pull down resistor is inserted between  $V_{OT}$  and  $V_{NC}$ (control GND), the extra circuit current, which is calculated approximately by  $V_{OT}$  output voltage divided by pull down resistance, flows as LVIC circuit current continuously. In the case of using  $V_{OT}$  for detecting high temperature over room temperature only, it is unnecessary to insert the pull down resistor.
- (3) In the case of using  $V_{OT}$  with low voltage controller like 3.3V MCU,  $V_{OT}$  output might exceed control supply voltage 3.3V when temperature rises excessively. If system uses low voltage controller, it is recommended to insert a clamp Di between control supply of the controller and  $V_{OT}$  output for preventing over voltage destruction.
- (4) In the case of not using  $V_{OT}$ , leave  $V_{OT}$  output NC (No Connection).

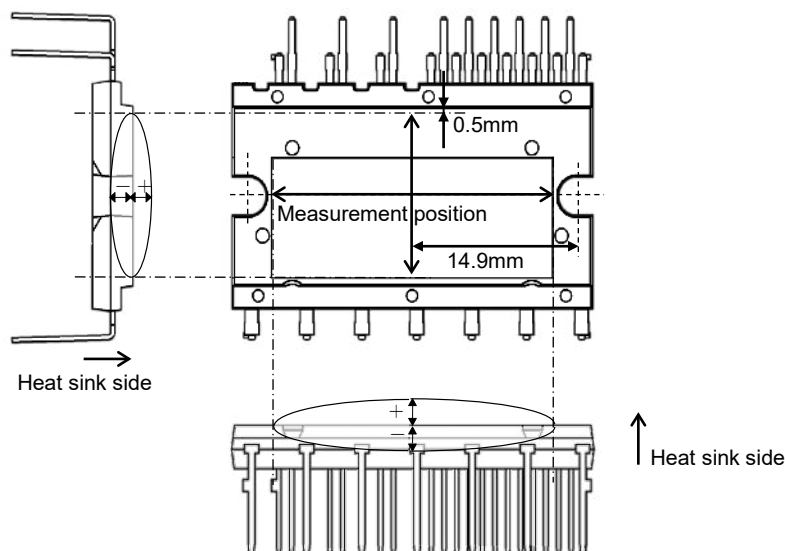
Refer the application note for SLIMDIP series about the usage of  $V_{OT}$ .

**MECHANICAL CHARACTERISTICS AND RATINGS**

| Parameter                    | Condition                                                             | Reference                       | Limits |      |      | Unit  |
|------------------------------|-----------------------------------------------------------------------|---------------------------------|--------|------|------|-------|
|                              |                                                                       |                                 | Min.   | Typ. | Max. |       |
| Mounting torque              | Mounting screw : M3 (Note 10)                                         | JEITA-ED-4701<br>402 method II  | 0.59   | 0.69 | 0.78 | N·m   |
| Terminal strength pull       | Control terminal: Load 5N<br>Power terminal: Load 10N                 | JEITA-ED-4701<br>401 method I   | 10     | -    | -    | s     |
| Terminal strength bending    | Control terminal: Load 2.5N<br>Power terminal: Load 5N<br>90deg. bend | JEITA-ED-4701<br>401 method III | 2      | -    | -    | times |
| Weight                       |                                                                       | -                               | -      | 5.5  | -    | g     |
| Heat radiation part flatness | (Note 11)                                                             | -                               | -30    | -    | 80   | μm    |

Note 10: Plain washers (ISO 7089~7094) are recommended.

Note 11: Measurement positions of heat radiation part flatness are as below

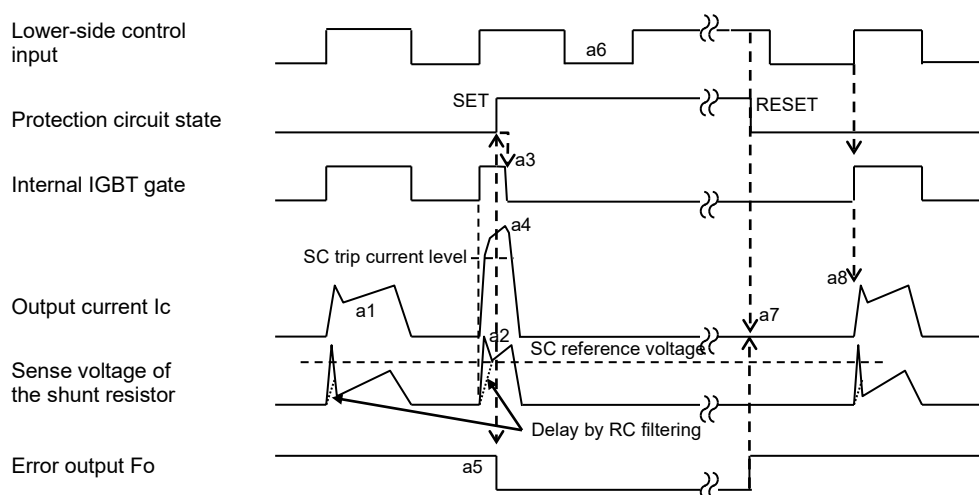
**RECOMMENDED OPERATION CONDITIONS**

| Symbol                         | Parameter                       | Condition                                                                 | Limits |      |      | Unit |
|--------------------------------|---------------------------------|---------------------------------------------------------------------------|--------|------|------|------|
|                                |                                 |                                                                           | Min.   | Typ. | Max. |      |
| $V_{CC}$                       | Supply voltage                  | Applied between P-NU, NV, NW                                              | 0      | 300  | 400  | V    |
| $V_D$                          | Control supply voltage          | Applied between $V_{P1}-V_{NC}$ , $V_{N1}-V_{NC}$                         | 13.5   | 15.0 | 16.5 | V    |
| $V_{DB}$                       | Control supply voltage          | Applied between $V_{UEB}-V_{UFS}$ , $V_{VFB}-V_{VFS}$ , $V_{WFB}-V_{WFS}$ | 13.0   | 15.0 | 18.5 | V    |
| $\Delta V_D$ , $\Delta V_{DB}$ | Control supply variation        |                                                                           | -1     | -    | +1   | V/μs |
| $t_{dead}$                     | Arm shoot-through blocking time | For each input signal                                                     | 1.0    | -    | -    | μs   |
| $f_{PWM}$                      | PWM input frequency             | $T_C \leq 115^\circ\text{C}$ , $T_J \leq 125^\circ\text{C}$               | -      | -    | 20   | kHz  |
| PWIN(on)                       | Minimum input pulse width       |                                                                           | 0.7    | -    | -    | μs   |
| PWIN(off)                      |                                 |                                                                           | 0.7    | -    | -    |      |
| $V_{NC}$                       | $V_{NC}$ variation              | Between $V_{NC}-NU$ , NV, NW (including surge)                            | -5.0   | -    | +5.0 | V    |
| $T_J$                          | Junction temperature            |                                                                           | -20    | -    | 125  | °C   |

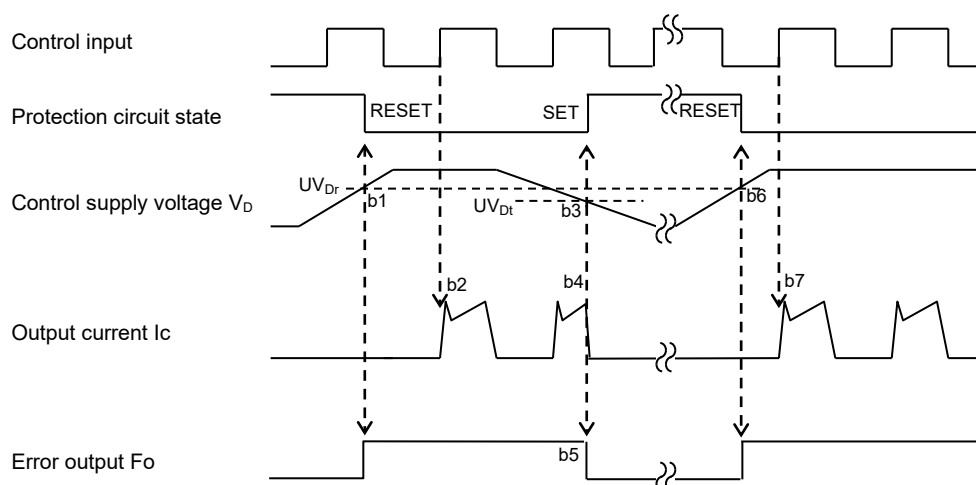
Fig. 5 Timing Charts of The DIIPM Protective Functions

**[A] Short-Circuit Protection (N-side only with the external shunt resistor and RC filter)**

- a1. Normal operation: IGBT ON and outputs current.
- a2. Short circuit current detection (SC trigger)  
(It is recommended to set RC time constant 1.5~2.0 $\mu$ s so that IGBT shut down within 2.0 $\mu$ s when SC.)
- a3. All N-side IGBT's gates are hard interrupted.
- a4. All N-side IGBTs turn OFF.
- a5.  $F_o$  outputs for  $t_{Fo}$ =minimum 20 $\mu$ s.
- a6. Input = "L": IGBT OFF
- a7.  $F_o$  finishes output, but IGBTs don't turn on until inputting next ON signal (L $\rightarrow$ H).  
(IGBT of each phase can return to normal state by inputting ON signal to each phase.)
- a8. Normal operation: IGBT ON and outputs current.

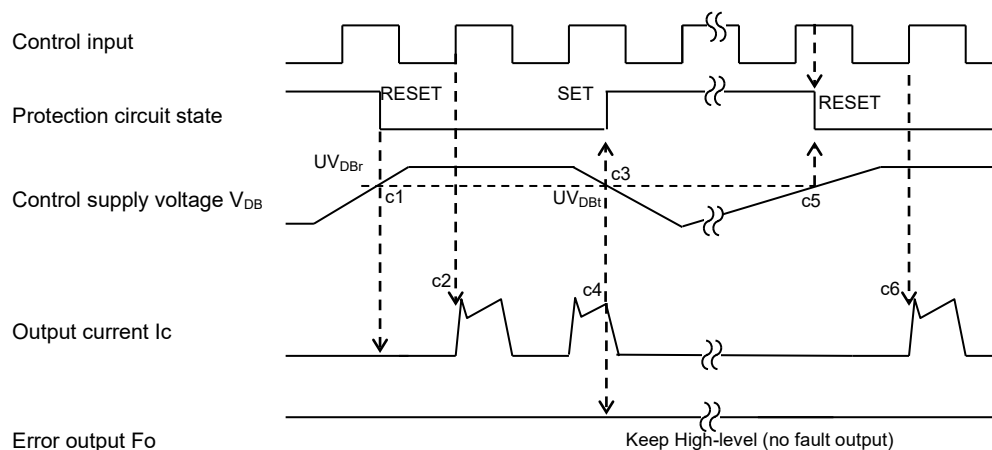
**[B] Under-Voltage Protection (N-side, UV<sub>D</sub>)**

- b1. Control supply voltage  $V_D$  exceeds under voltage reset level ( $UV_{Dr}$ ), but IGBT turns ON by next ON signal (L $\rightarrow$ H).  
(IGBT of each phase can return to normal state by inputting ON signal to each phase.)
- b2. Normal operation: IGBT ON and outputs current.
- b3.  $V_D$  level drops to under voltage trip level. ( $UV_{Dt}$ ).
- b4. All N-side IGBTs turn OFF in spite of control input condition.
- b5.  $F_o$  outputs for  $t_{Fo}$ =minimum 20 $\mu$ s, but output is extended during  $V_D$  keeps below  $UV_{Dr}$ .
- b6.  $V_D$  level reaches  $UV_{Dr}$ .
- b7. Normal operation: IGBT ON and outputs current.

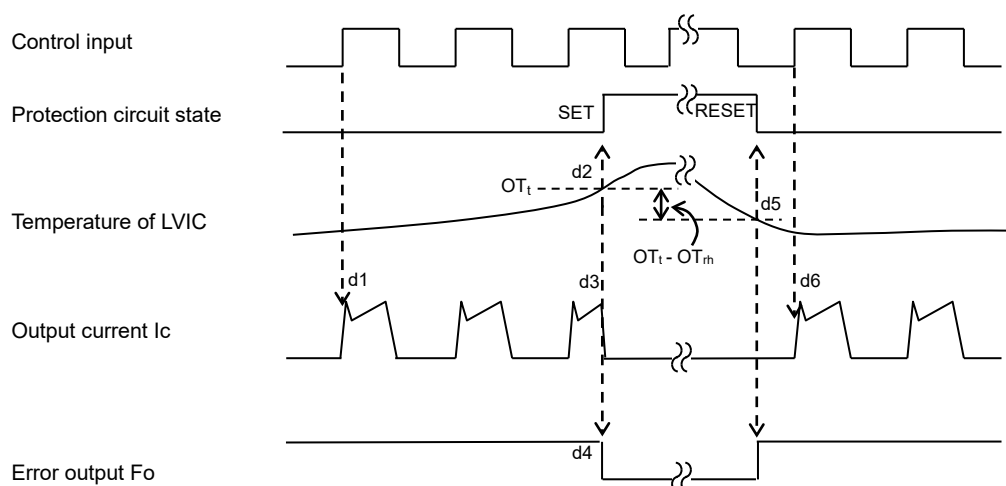


**[C] Under-Voltage Protection (P-side,  $UV_{DB}$ )**

- c1. Control supply voltage  $V_{DB}$  rises. After the voltage reaches under voltage reset level  $UV_{DBr}$ , IGBT turns on by next ON signal (L→H).  
 c2. Normal operation: IGBT ON and outputs current.  
 c3.  $V_{DB}$  level drops to under voltage trip level ( $UV_{DBt}$ ).  
 c4. IGBT of the corresponding phase only turns OFF in spite of control input signal level, but there is no  $F_o$  signal output.  
 c5.  $V_{DB}$  level reaches  $UV_{DBr}$ .  
 c6. Normal operation: IGBT ON and outputs current.

**[D] Over Temperature Protection (N-side, Detecting LVIC temperature)**

- d1. Normal operation: IGBT ON and outputs current.  
 d2. LVIC temperature exceeds over temperature trip level ( $OT_t$ ).  
 d3. All N-side IGBTs turn OFF in spite of control input condition.  
 d4.  $F_o$  outputs for  $t_{Fo}$ =minimum 20 $\mu$ s, but output is extended during LVIC temperature keeps over  $OT_t$ .  
 d5. LVIC temperature drops to over temperature reset level.  
 d6. Normal operation: IGBT turns on by next ON signal (L→H).  
 (IGBT of each phase can return to normal state by inputting ON signal to each phase.)



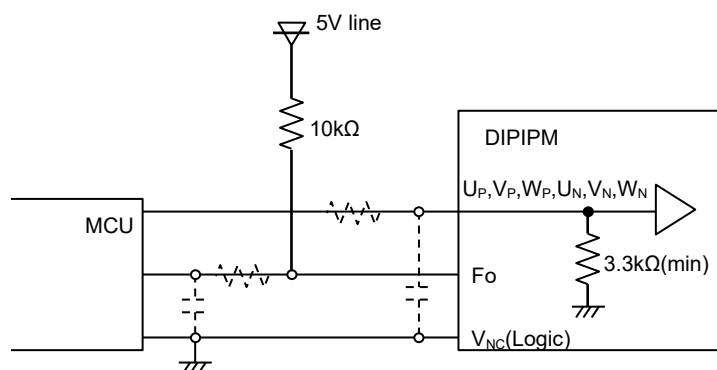
The diagram illustrates a motor driver circuit. On the left, an MCU is connected to an HVIC (High Voltage Integrated Circuit) and an LVIC (Low Voltage Integrated Circuit). The HVIC controls six RC-IGBTs (RC-Insulated Gate Bipolar Transistors) which drive a motor (M). The LVIC controls the base of the RC-IGBTs. The circuit includes various pins for power, ground, and control signals. Key components include capacitors (C1, C2, C3, C4), resistors (R1, R2), and a shunt resistor (N1). Callouts provide additional information:

- MCU:** In the case of being affected by noise, it is recommended to insert RC filter.
- Long GND wiring:** Long GND wiring might generate noise to input signal and cause IGBT malfunction.
- Long wiring:** Long wiring might cause SC level fluctuation and malfunction.
- Shunt resistor:** Long wiring might cause short circuit failure.

- 8

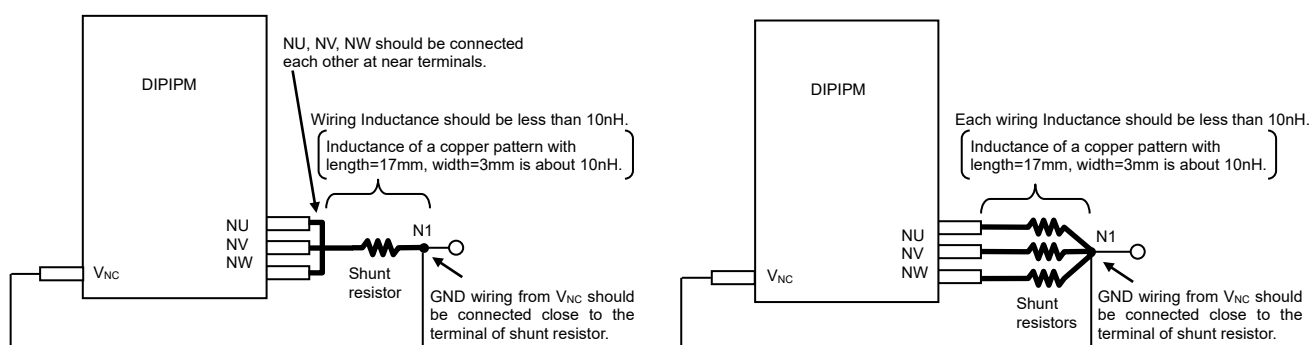


Fig. 7 MCU I/O Interface Circuit



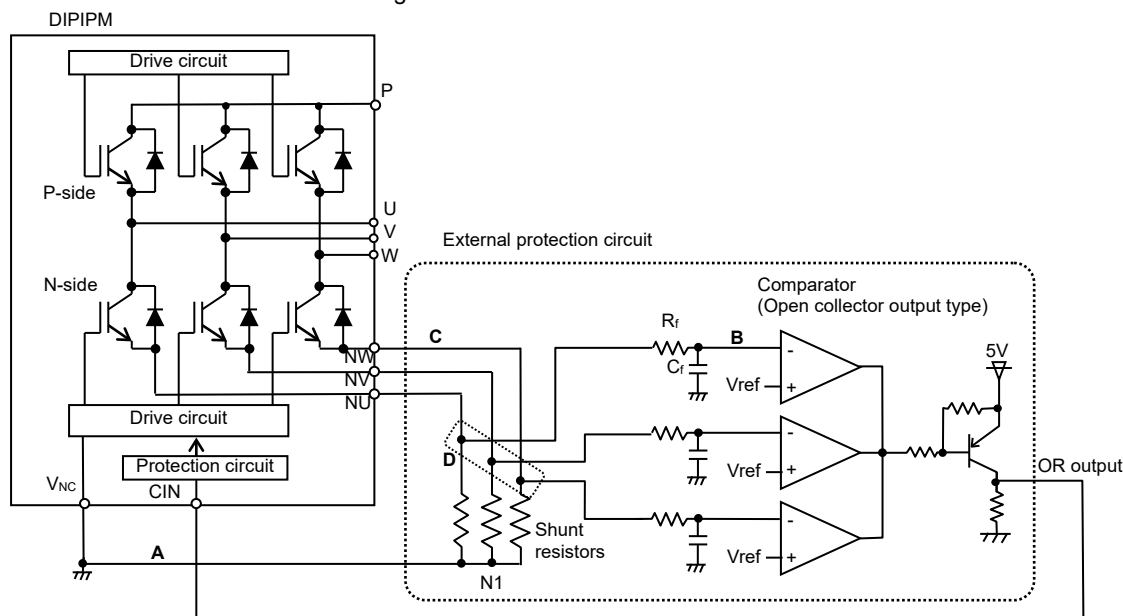
Note: The RC coupling (parts shown in the dotted line) at each input depends on user's PWM control strategy and the wiring impedance of the printed circuit board. The DIIPM signal input section integrates a 3.3kΩ(min) pull-down resistor. Therefore, when using an external filtering resistor, please pay attention to the signal voltage drop at input terminal.

Fig. 8 Pattern Wiring Around the Shunt Resistor



Low inductance shunt resistor like surface mounted (SMD) type is recommended.

Fig. 9 External SC Protection Circuit with Using Three Shunt Resistors



- (1) It is necessary to set the time constant  $R_r C_r$  of external comparator input so that IGBT stop within  $2\mu s$  when short circuit occurs. SC interrupting time might vary with the wiring pattern, comparator speed and so on.
- (2) The threshold voltage  $V_{ref}$  should be set up the same rating of short circuit trip level ( $V_{sc(ref)}$  typ. 0.48V).
- (3) Select the external shunt resistance so that SC trip-level is less than specified maximum value.
- (4) To avoid malfunction, the wiring A, B, C should be as short as possible.
- (5) The point D at which the wiring to comparator is divided should be near the terminal of shunt resistor.
- (6) OR output high level should be over 0.505V (=maximum  $V_{sc(ref)}$ ).
- (7) GND of Comparator,  $V_{ref}$  circuit and  $C_r$  should be not connected to noisy power GND but to control GND wiring.

**SLIMDIP-M**

TRANSFER MOLDING TYPE  
INSULATED TYPE

Fig. 10 (Reference Figure) PCB Through-hole Pattern

Recommended through-hole locations and diameter layout for SLIMDIP

[Dimension: mm]

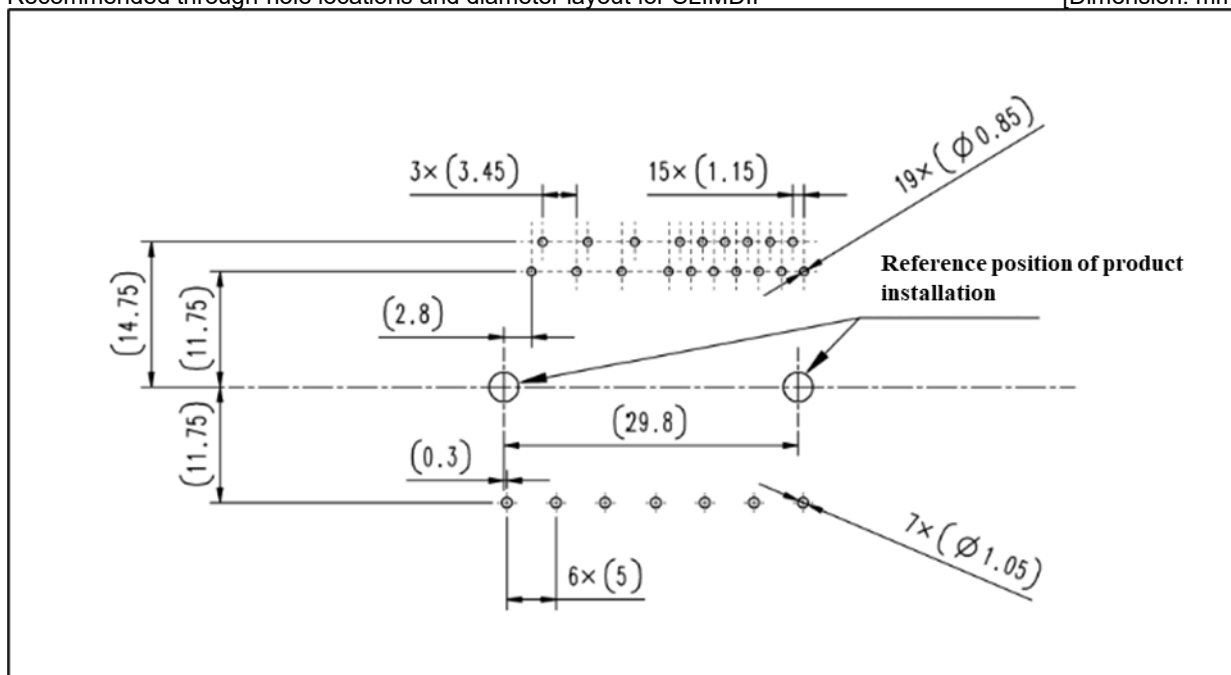
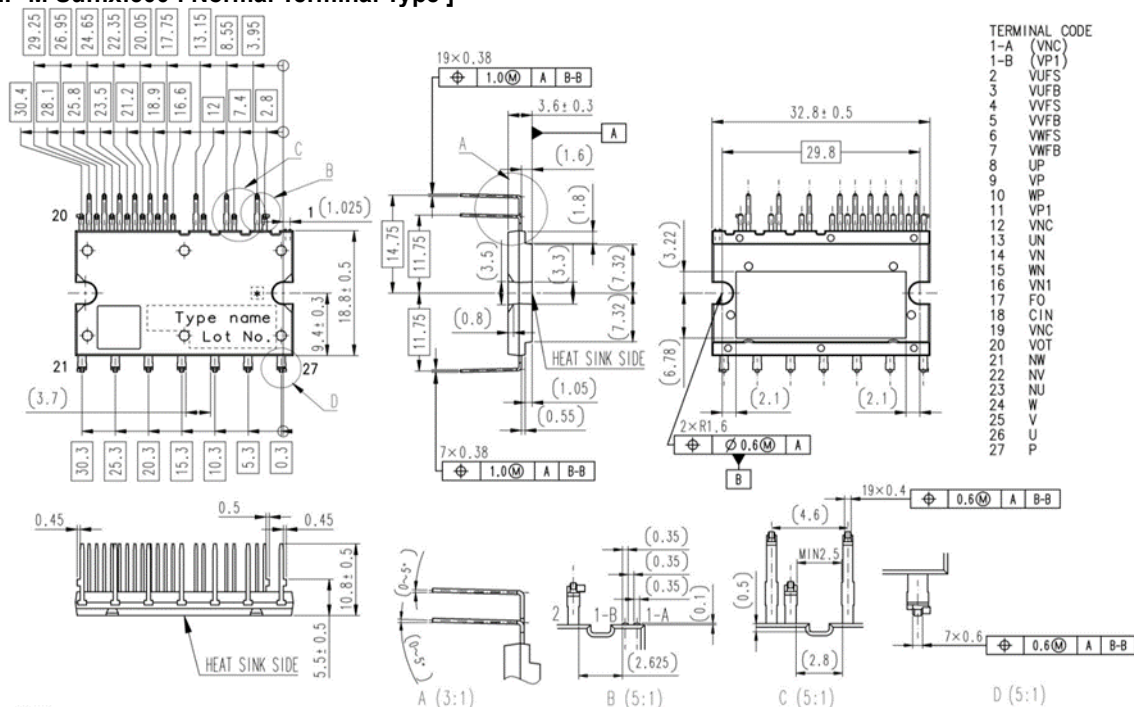
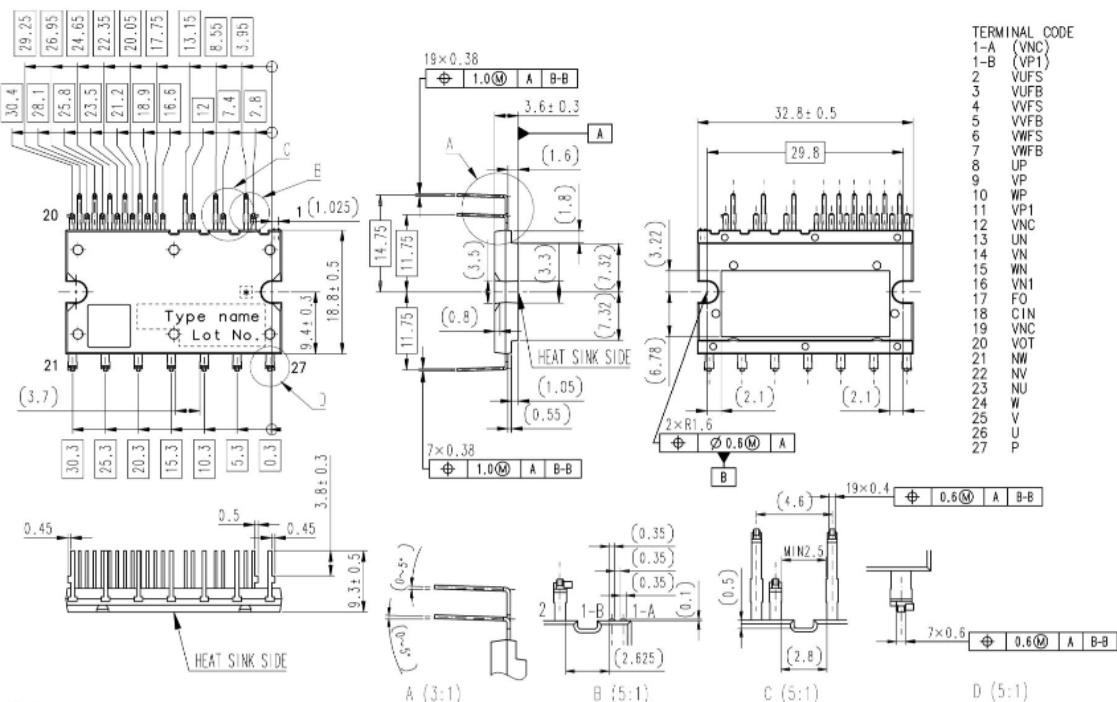


Fig. 11 Package Outlines

[Dimension: mm]

**[ SLIMDIP-M Suffix:500 : Normal Terminal Type ]****[ SLIMDIP-M Suffix:505 : Short Terminal Type ]**

Note: Connect only one V<sub>NC</sub> terminal (No.12 or 19) to the system GND and leave another one open.

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